

1 Introduction

1.1 Angular Spectrum of Plane Waves

A Gaussian beam with a beam waist of w_0 has a corresponding angular representation

$$A_0(\theta, \phi) = \frac{1}{N} e^{-\left(\frac{\theta}{\theta_0}\right)^2} \quad (1)$$

where

$$\theta_0 = \frac{\lambda}{\pi w_0} \quad (2)$$

and the normalization N is chosen so that

$$\int_{\phi=0}^{2\pi} \int_{\theta=0}^{\pi/2} A_0(\theta, \phi)^2 \sin \theta \, d\theta \, d\phi = 1 \quad (3)$$

so that the coupling integral, I_{00} , for any beam with itself is unity

$$I_{ij} = \int_{\phi=0}^{2\pi} \int_{\theta=0}^{\pi/2} A_i^*(\theta, \phi) A_j(\theta, \phi) \sin \theta \, d\theta \, d\phi \quad (4)$$

1.2 Propagation of the Beam

Each infinite plane wave, propagating through a distance z in a vacuum will undergo a phase shift $\vec{k} \cdot \vec{z}$. The following equations were implemented in the Octave function `propagatebeam.m`.

$$A'(\theta, \phi) = A(\theta, \phi) e^{ikz \cos \theta} \quad (5)$$

$$k = \frac{2\pi\nu}{c} \quad (6)$$

For propagation through layers of material the Octave function `layers_complex.m` was used. This returns amplitude transmission coefficients t_{TE} and t_{TM} for a stack of layers with given complex values of ε and μ . The TE and TM components of each plane wave were identified as $A(\theta, \phi) \cos \phi$ and $A(\theta, \phi) \sin \phi$ respectively. Then using the same relations to re-assemble the beam, propagation through the layers is given by:

$$\begin{aligned} A'(\theta, \phi) &= \cos^2(\phi) t_{\text{TE}}(\theta, z, \nu) A(\theta, \phi) \\ &+ \sin^2(\phi) t_{\text{TM}}(\theta, z, \nu) A(\theta, \phi) \end{aligned} \quad (7)$$

The Octave function `layersbeam.m` performs this calculation.

2 Test Runs

A number of test runs were performed as a function of layer thickness, z . In each case a set of beam coefficients, A_0 , for a given angle, θ_0 , was generated. The beam was the propagated both using the `layers_complex.m` code and also the simple beam propagation function for the same distance.

$$A_1 = \text{layersbeam.m } \nu, \varepsilon, z, A_0 \quad (8)$$

$$A_2 = \text{propagatebeam.m } \nu, z, A_0 \quad (9)$$

Finally the magnitude of the coupling integral was taken

$$|I_{12}| \quad (10)$$

which should be equal to 1 for an plane wave in air due to the correction for distance applied to A_2 .

To test the routines, this procedure was carried out for vacuum ($\varepsilon = 1$), giving the result shown in [Figure 1](#). As expected this gives a value of $I_{12} = 1$ for each distance and beam angle.

The procedure was also tested for values of $\varepsilon = 2, 4$ ($n = \sqrt{2}, 2$) giving the results shown in [Figure 2](#) and [Figure 3](#).

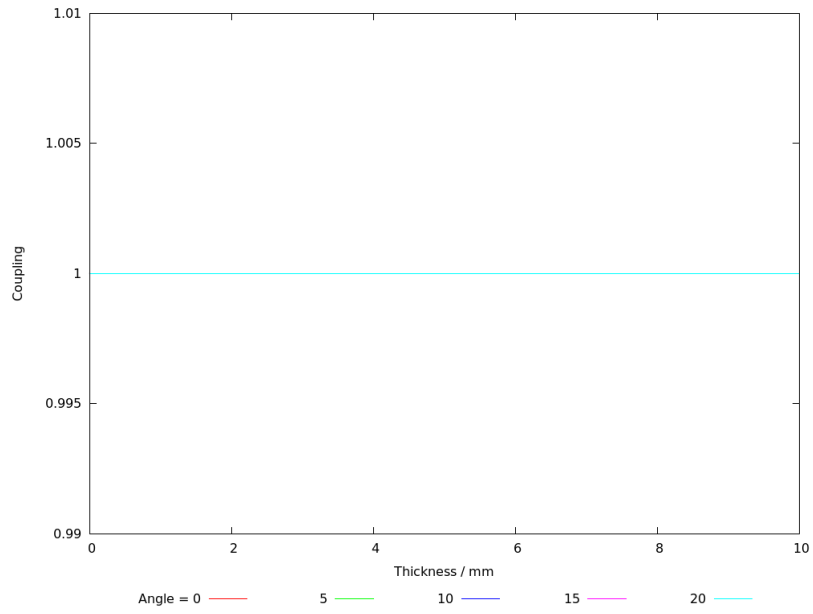


Figure 1: $\varepsilon = 1$

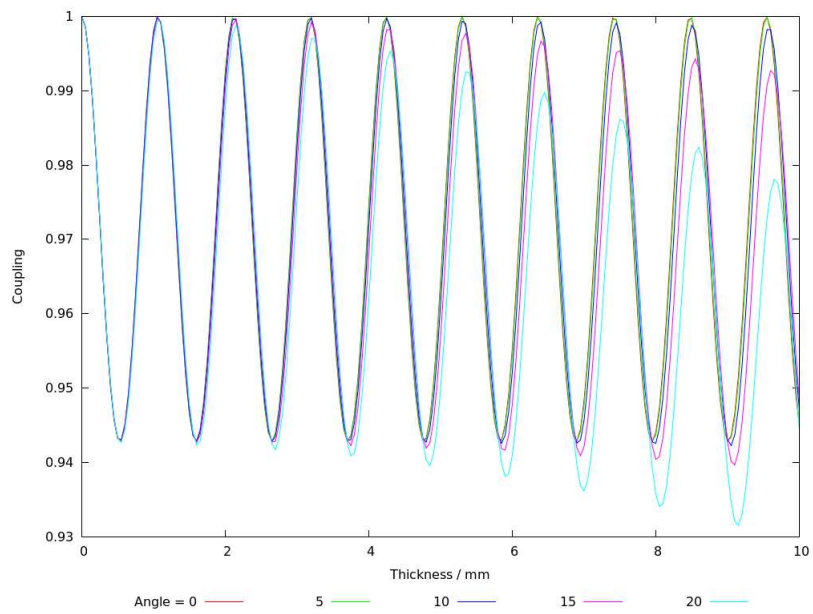


Figure 2: $\varepsilon = 2$

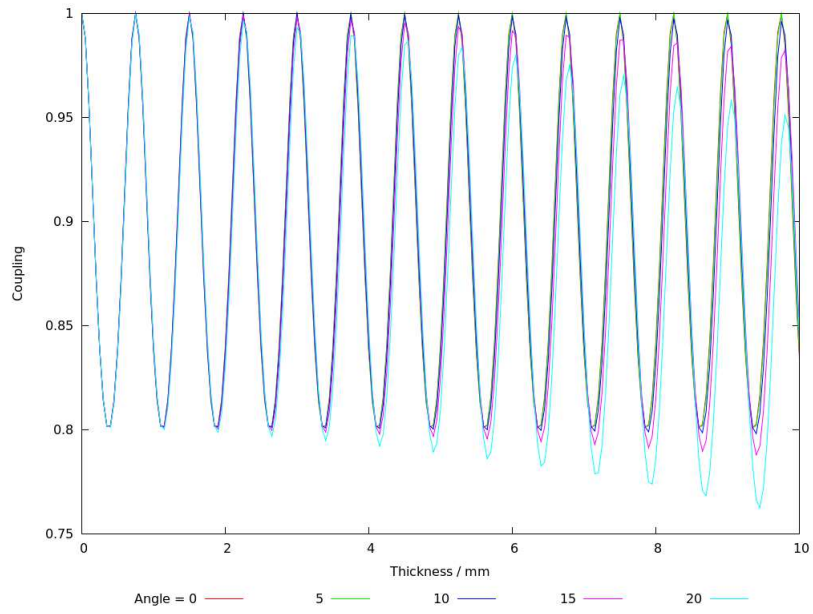


Figure 3: $\varepsilon = 4$

3 Results

The routine was then run as a function of frequency with the following parameters:

- Refractive index $n = 2$ ($\varepsilon = 4$)
- Thickness $z = 10$ mm
- Frequency $\nu = 75 - 110$ GHz

The results are plotted in [Figure 4](#). Finally the period of the oscillations was fitted in order to determine the refractive index of the material, giving the plot shown in [Figure 5](#).

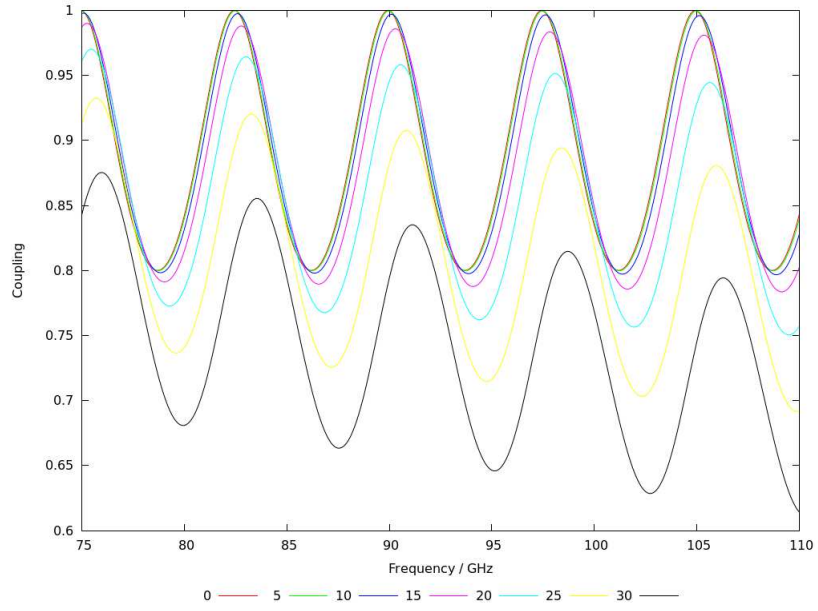


Figure 4: $\varepsilon = 4$

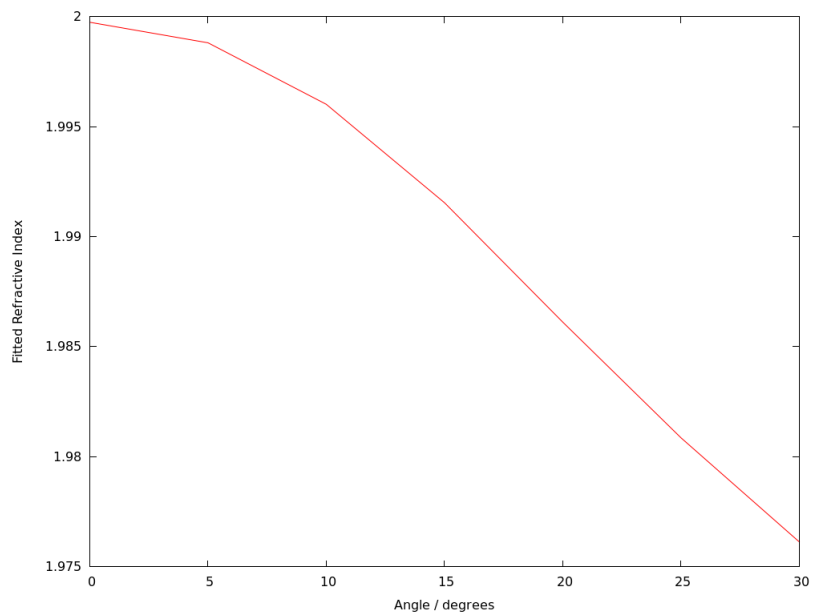


Figure 5: Fitted refractive index for each angle θ_0 .

A Numerical Evaluation of Coupling Integral

A.1 Evaluation as a Sum

Sample points are defined as follows, where in order to handle the case of a pure plane wave $\theta = 0$ represents the range $0 \leq \theta < \Delta\theta/2$ and other values represent the range $\theta_i - \Delta\theta/2 \leq \theta < \theta_i + \Delta\theta/2$.

$$\theta : 0 \quad \dots \quad (n_\theta - 1) \frac{\pi}{2n_\theta - 1}, \quad \Delta\theta = \frac{\pi}{2n_\theta - 1} \quad (11)$$

$$\phi : -\pi(1 - \frac{2}{n_\phi}) \quad \dots \quad \pi, \quad \Delta\phi = \frac{2\pi}{n_\phi} \quad (12)$$

Then the surface area of the corresponding element unit sphere is given by

$$\begin{aligned} \int_{\phi - \Delta\phi/2}^{\phi + \Delta\phi/2} \int_{\theta - \Delta\theta/2}^{\theta + \Delta\theta/2} \sin \theta \, d\theta \, d\phi &= \left[\phi \left[-\cos \theta \right]_{\theta - \Delta\theta/2}^{\theta + \Delta\theta/2} \right]_{\phi - \Delta\phi/2}^{\phi + \Delta\phi/2} \\ &= \Delta\phi (\cos(\theta - \Delta\theta/2) - \cos(\theta + \Delta\theta/2)) \\ &= 2 \sin \theta \sin(\Delta\theta/2) \Delta\phi \end{aligned} \quad (13)$$

or in the case of $\theta = 0$

$$\left[\phi \left[-\cos \theta \right]_0^{\Delta\theta/2} \right]_{\phi - \Delta\phi/2}^{\phi + \Delta\phi/2} = (1 - \cos(\Delta\theta/2)) \Delta\phi$$

So the final expression for discrete sample points is:

$$I_{ij} = \sum_{\phi=0}^{2\pi} \sum_{\theta=0}^{\pi/2} A_i^*(\theta, \phi) A_j(\theta, \phi) \begin{cases} (1 - \cos(\Delta\theta/2)) \Delta\phi & \theta = 0 \\ 2 \sin \theta \sin(\Delta\theta/2) \Delta\phi & \text{otherwise} \end{cases} \quad (14)$$

A.2 Minimum Number of Sample Points

To ensure a smooth variation of phase between the sample points we can impose a limit

$$|kz \cos \theta_{i+1} - kz \cos \theta_i| \ll \frac{\pi}{8} \quad (15)$$

which can be rearranged to

$$\Delta\theta \sin \theta \ll \frac{c}{16\nu z} \quad (16)$$

or taking the worst case ($\sin \theta = 1$) and a general medium

$$\Delta\theta \ll \frac{c}{16\nu z \sqrt{\epsilon\mu}} \quad (17)$$

For example with $n = 2$, $\nu = 100$ GHz, $z = 1$ mm

$$\Delta\theta \ll 5^\circ \quad \Rightarrow \quad \gg 18 \text{ points} \quad (18)$$